#### **DESCRIPTION**

The AO321K consist of single independent high-gain frequency-compensated operational amplifier that is designed specifically to operate from a single supply or split supply over a wide range of voltages.

The AO321K have a high gain-bandwidth product of 1MHz, a slew rate of  $0.2V/\mu s$ , and a quiescent current of  $250\mu A/amplifier$  at 5V.

The AO321K is designed to provide optimal performance in low voltage and low noise systems.

The maximum input offset voltage is 5mV for AO321K. The operating range is from 3V to 36V.

The AO321K Quad is available in SOT-25 package.

#### ORDERING INFORMATION

Package Type	Part Number			
SOT-25	E5	AO321KE5R		
SPQ: 3,000pcs/Reel	Lo	AO321KE5VR		
Note	V: Halogen free Package R: Tape & Reel			
AiT provides all RoHS products				

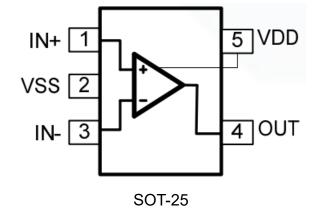
## **FEATURES**

- Single-Supply Operation from +3V ~ +36V
- Dual-Supply Operation from ±1.5V ~ ±18V
- Gain-Bandwidth Product: 1MHz (Typ)
- Low Input Bias Current: 20nA (Typ)
- Low Offset Voltage: 5mV (Max)
- Quiescent Current: 250µA per Amplifier (Typ)
- Input Common Mode Voltage Range Includes Ground
- Operating Temperature: -25°C ~ +85°C

# **APPLICATION**

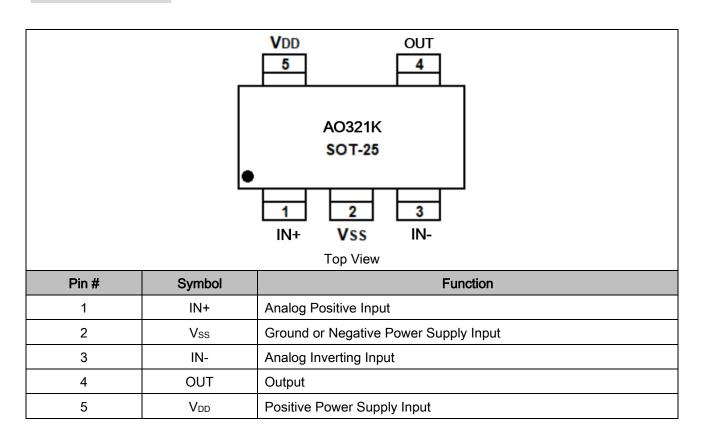
- Home Theaters, Chemical and Gas Sensors
- Digital Multimeter: Bench and Systems
- Field Transmitter: Temperature Sensors
- Mother Control: AC Induction, Brushed DC, Brushless DC, High-Voltage, Low-Voltage, Permanent Magnet, and Stepper Motor
- Oscilloscopes
- TV: LED and Digital
- Temperature Sensors and Controllers Using Modbus
- Weight Scales
- Walkie-Talkie, Multivibrators
- Battery Management Solution
- Transducer Amplifiers, Summing Amplifiers
- Portable Systems

## TYPICAL APPLICATION



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# PIN DESCRIPTION



# **ABSOLUTE MAXIMUM RATINGS**

Parameter	Symbol	Max
Power Supply Voltage	Vcc	±20V or 40V
Differential input voltage	$V_{I(DIFF)}$	40V
Input Voltage	Vı	-0.3V~40V
Operating Temperature Range	Topr	-25°C ~+85°C
Storage Temperature Range	Tstg	-65°C ~+150°C

Stress beyond above listed "Absolute Maximum Ratings" may lead permanent damage to the device. These are stress ratings only and operations of the device at these or any other conditions beyond those indicated in the operational sections of the specifications are not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

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# **ELECTRICAL CHARACTERISTICS**

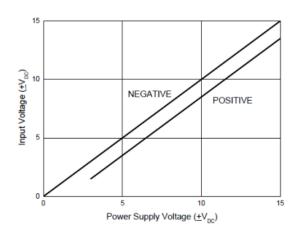
V<sub>S</sub> = +15V, T<sub>A</sub>=25°C, unless otherwise noted.

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
INPUT CHARACTERISTICS						
Input Offset Voltage	Vos	V <sub>CM</sub> = V <sub>S</sub> /2	1	0.4	5	mV
Input Bias Current	lΒ		-	20	-	nA
Input Offset Current	los		-	5	-	nA
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>S</sub> = 5.5V	-	-0.1 to +4	-	V
Common-Mode Rejection Ratio	CMRR	V <sub>CM</sub> = 0V to Vs-1.5V	60	70	-	dB
Open-Loop Voltage Gain	Aol	$R_L = 5k\Omega$ , $V_O = 1V$ to 11V	85	100	-	dB
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$		-	7	-	μV/°C
OUTPUT CHARACTERISTICS						
Output Voltage Swing from Rail	Vон	$R_L = 2k\Omega$	-	11	-	V
	$V_{OL}$	$R_L = 2k\Omega$	1	5	20	mV
	Vон	$R_L = 10k\Omega$	13-	12	-	V
	Vol	$R_L = 10k\Omega$	1	5	20	mV
Output Current	I <sub>SOURCE</sub>	$R_L = 10\Omega$ to $V_S/2$	ı	40	60	- mA
	Isink		-	40	60	
POWER SUPPLY						
Operating Voltage Range			3	-	-	V
			-	-	36	V
Power Supply Rejection Ratio	PSRR	$V_S = +5V \text{ to } +30V,$ $V_{CM} = +0.5V$	70	100	-	dB
Quiescent Current/Amplifier	la	Vs=30V, R <sub>L</sub> =∞	-	0.25	2.0	mA
DYNAMIC PERFORMANCE						
Gain-Bandwidth Product	GBP		ı	1	-	MHz
Slew Rate	SR	G = +1, 2V Output Step	-	0.2	-	V/µs

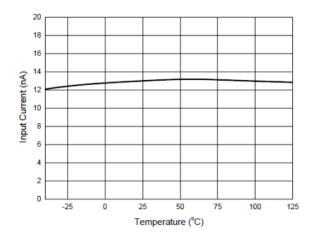
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# TYPICAL PERFORMANCE CHARACTERISTICS

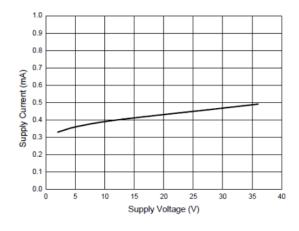
#### 1. Input Voltage Range



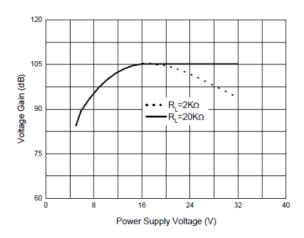
#### 2. Input Current



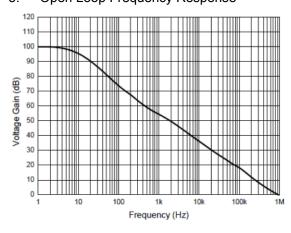
## 3. Supply Current



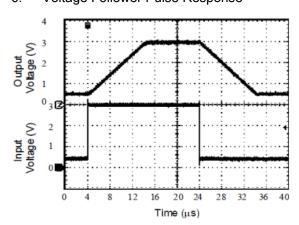
4. Voltage Gain



## 5. Open Loop Frequency Response

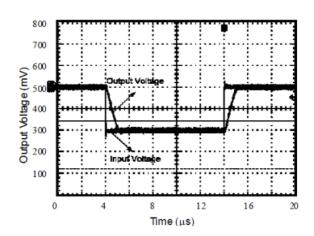


6. Voltage Follower Pulse Response

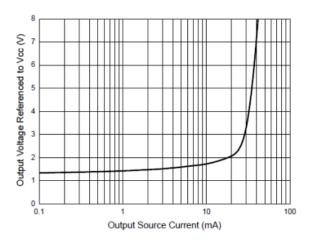


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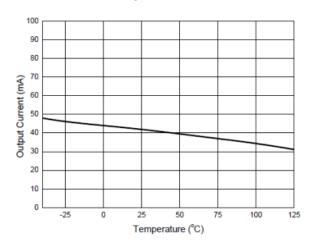
#### 7. Voltage Follower Pulse Response(Small Signal)



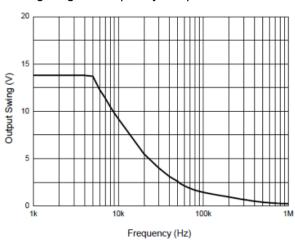
# 9. Output Characteristics: Current Sourcing



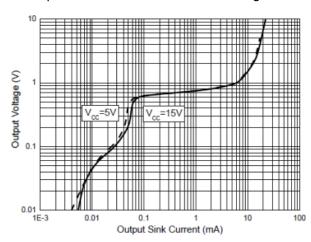
#### 11. Current Limiting



#### 8. Large Signal Frequency Response

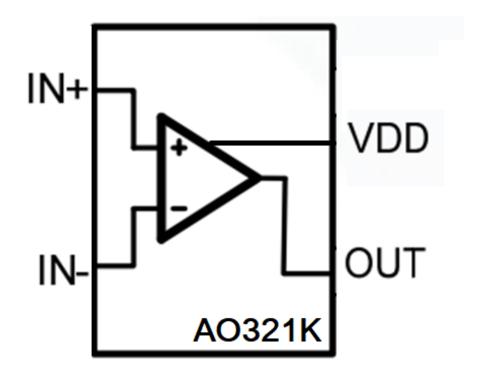


## 10. Output Characteristics: Current Sinking



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# **BLOCK DIAGRAM**



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## **DETAILED INFORMATION**

AO321K op amps are unity-gain stable and suitable for a wide range of general-purpose applications.

#### Power Supply Bypassing and Board Layout

AO321K operates from a single 3V to 36V supply or dual  $\pm 1.5$ V to  $\pm 18$ V supplies. For best performance, a 0.1 $\mu$ F ceramic capacitor should be placed close to the V<sub>DD</sub> pin in single supply operation. For dual supply operation, both V<sub>DD</sub> and V<sub>SS</sub> supplies should be bypassed to ground with separate 0.1 $\mu$ F ceramic capacitors.

#### **Low Supply Current**

The low supply current (typical 250uA per channel) of AO321K will help to maximize battery life.

#### **Operating Voltage**

AO321K operates under wide input supply voltage (3V to 36V). In addition, all temperature specifications apply from -25°C to +85°C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime.

#### Capacitive Load Tolerance

The AO321K is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain.

Figure 1. shown a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

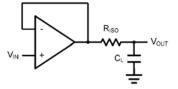


Figure 1. Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 2 is an improvement to the one in Figure 1.  $R_F$  provides the DC accuracy by feed-forward the  $v_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

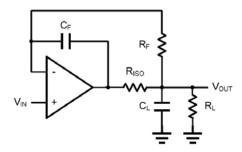


Figure 2. Indirectly Driving a Capacitive Load with DC Accuracy

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## **Typical Application Circuits**

#### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 3 shown the differential amplifier using AO321K.

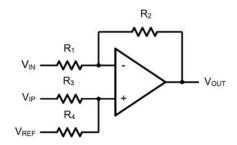


Figure 3. Differential Amplifier

$$V_{OUT} = \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_4}{R_1} V_{IN} - \frac{R_2}{R_1} V_{IP} + \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal (i.e. R<sub>1</sub>=R<sub>3</sub> and R<sub>2</sub>=R<sub>4</sub>), then

$$V_{OUT} = \frac{R_2}{R_1} (V_{IP} - V_{IN}) + V_{REF}$$

#### Low Pass Active Filter

The low pass active filter is shown in Figure 4. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

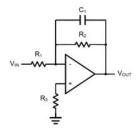


Figure 4. Low Pass Active Filter

## **Instrumentation Amplifier**

The triple AO321K can be used to build a three-op-amp instrumentation amplifier as shown in Figure 5. The amplifier in Figure 5 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

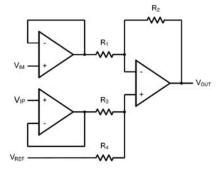
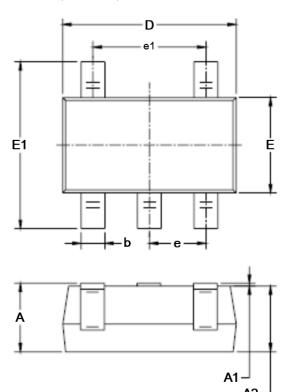


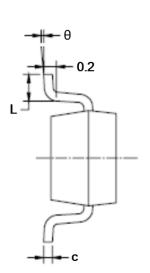
Figure 5. Instrument Amplifier

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# PACKAGE INFORMATION

Dimension in SOT-25 (Unit: mm)





Complete al	Millimeters			
Symbol	Min	Max		
А	1.050	1.250		
A1	0.000 0.100			
A2	1.050 1.150			
b	0.300	0.500		
С	0.100	0.200		
D	2.820	3.020		
Е	1.500	1.700		
E1	2.650	2.950		
е	0.950BSC			
e1	1.900BSC			
L	0.300 0.600			
θ	0°	8°		

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